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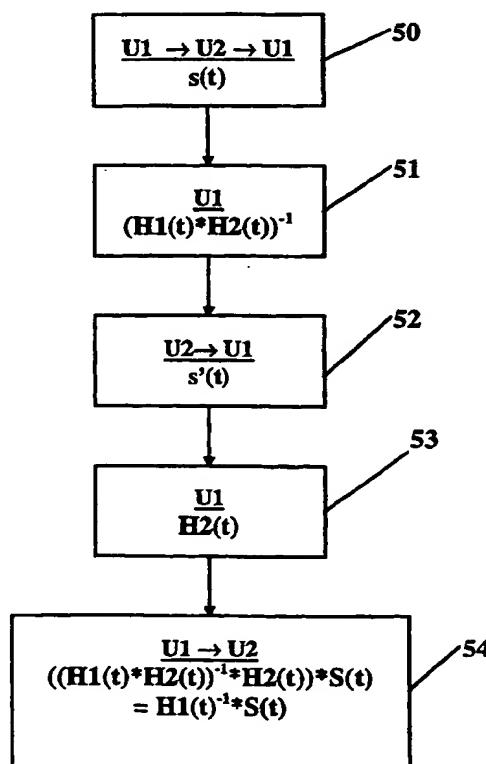
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(54) Method for pre-distorting signals transmitted over non-reciprocal channels

(57) The present invention refers to a method of predistorting a signal sent from a first unit (U1) to a second unit, the said first unit (U1) sending data to the said second unit (U2) through a first channel and the said second unit (U2) sending data to the said first unit (U1) through a second channel. The method comprises the computing (53) of representative coefficients of the second channel (H2) on the basis of a second predetermined sequence (s'(t)) sent (52) from the second unit (U2) to the first unit (U1); and it is characterized in that it comprises the following steps:

- the sending (50) by the first unit (U1) of a first predetermined sequence (s(t)) to the second unit (U2) which sends it back in the form received to the first unit (U1), in the form of a return sequence;
- in the said first unit (U1), the computing (51) of representative coefficients ($\{H_1 \cdot H_2\}^{-1}$, $\{H_1 \cdot H_2\}$) of the first and second channels jointly (H_1, H_2) on the basis of the return sequence received (s(t)); and
- predistortion of the signal sent by the first unit according to the said representative coefficients of the second channel (H_2, H_2^{-1}) and the said representative coefficients of the first and second channels jointly ($\{H_1 \cdot H_2\}^{-1}, \{H_1 \cdot H_2\}$).

FIG 6



Description**OBJECT OF THE INVENTION**

[0001] The present invention refers to a method for predistortion of a signal transmitted or sent from a first unit to a second unit of a telecommunications system, as well as a unit for carrying out the method. For example, the two units exchange data or voice signals through a radio channel, an optical fibre, a copper wire, or any other type of physical medium.

STATE OF THE ART

[0002] Typically, a unit in accordance with the present invention equalizes the received signal and predistorts the transmitted signal. The unit comprises an equalizer in the reception chain and a predistorter in the transmission chain. The object of incorporating the equalization and predistortion functions in the same unit is to reduce the cost of the other remote unit/s exchanging data with the said unit; these remote units do not need equalizers.

[0003] A unit of this type, including an equalizer and a predistorter, is known from the document "Adaptive channel precoding for personal communications" by W. ZHUANG et al., published in the magazine "Electronics Letters" on 15th September 1994, Vol. 30, No. 19, pages 1570 - 1571. In accordance with this document, over a period of time, the same equalization coefficients computed in reception are used to predistort the signal transmitted. This solution results from the hypothesis that the interferences or distortions that are applied to the sent or transmitted signal are the same as the interferences that are applied to the signal received, that is, that the characteristics of the transmission and reception channels are the same.

[0004] This hypothesis is false if the respective transmission and reception channels are different, for example (a) - through being defined by respective frequencies that are different among themselves, as is the case in Frequency Duplex Division (FDD), or - (b) - through being defined by physical media that are different among themselves, etc...

CHARACTERIZATION OF THE INVENTION

[0005] One object of the present invention is to define methods of predistorting a signal transmitted from a first unit to a second unit in a telecommunications system, taking into account the difference in characteristics between the respective transmission and reception channels. Accordingly, a method of predistorting a signal sent from a first unit to a second unit, the first unit sending data to the second unit through a first channel and the second unit sending data to the first unit through a second channel, the method comprising the following step:

- in the first unit, the computing of representative coefficients of the second channel on the basis of a second predetermined sequence sent from the second unit to the first unit; is characterized in that it comprises the following steps:
 - the sending by the first unit of a first predetermined sequence to the second unit which sends it back in the form received to the first unit, in the form of a return sequence;
 - in the said first unit, the computing of representative coefficients of the first and second channels jointly, on the basis of the return sequence received; and
 - predistortion of the signal sent by the first unit by means of inverse coefficients of the first channel which are computed according to the said representative coefficients of the second channel and the said representative coefficients of the first and second channels jointly.

[0006] According to a first sub-alternative, the representative coefficients of the second channel are inverse coefficients of the second channel, and the representative coefficients of the first and second channels jointly are direct coefficients of the first and second channels jointly, predistortion being carried out with the inverse coefficients of the convolution of both the inverse coefficients of the second channel and the direct coefficients of the first and second channels jointly.

[0007] According to another sub-alternative, the representative coefficients of the second channel are direct coefficients of the second channel and the representative coefficients of the first and second channels jointly are inverse coefficients of the first and second channels jointly, predistortion being carried out by means of the coefficients of the convolution of the direct coefficients of the second channel and of the inverse coefficients of the first and second channels.

[0008] Another object of the present invention is to define units for predistorting a signal to be transmitted in accordance with the methods defined above. Accordingly, a unit for carrying out this method comprises:

- means to predistort the signal to be sent; and it is characterized in that it comprises:
 - means to compute representative coefficients of the second channel on the basis of a second predetermined sequence sent from the remote unit to the unit;
 - means to send a first predetermined sequence to the remote unit, which sends it back in the form received to the said unit, in the form of a return sequence;
 - means to compute representative coefficients of the first and second channels jointly, on the basis of the return sequence received; and in that
 - the predistorting means predistort the signal sent by the inverse coefficients of the first channel which

are computed according to the said representative coefficients of the second channel and the said representative coefficients of the first and second channels jointly.

[0009] In accordance with the present invention, another method to predistort a signal sent from a first unit to a second unit, the first unit sending data to the said second unit through a first channel and the second unit sending data to the first unit through a second channel, the method comprising the following step:

- in the first unit, the computing of representative coefficients of the second channel on the basis of a second predetermined sequence sent from the second unit to the first unit; and it is characterized in that it comprises the following steps:
- the sending by the first unit to the second unit of a first sequence which results from the convolution of a predetermined sequence and of the said representative coefficients of the second channel, the first sequence being sent back in the form it was received by the second unit to the first unit in the form of a return sequence;
- in the said first unit, the computing of inverse coefficients of the first channel on the basis of the said return sequence; and
- predistortion of the signal sent by the first unit on the basis of the said inverse coefficients of the first channel.

BRIEF EXPLANATION OF THE DRAWINGS

[0010] A more detailed explanation showing other characteristics and advantages of the present invention is given in the following specification based on the attached drawings, in which:

- Figure 1 shows two units exchanging data through two respective channels which are different from each other;
- Figure 2 shows a block diagram of a first unit of Figure 1;
- Figure 3 shows a block diagram of a second unit of Figure 1;
- Figure 4 is a block diagram of a predistorter included in the first unit of Figure 2;
- Figure 5 is a block diagram of a generator of predistortion coefficients included in the first unit of Figure 2;
- Figure 6 shows a predistortion algorithm of a signal transmitted by the first unit according to a first embodiment of the invention; and
- Figure 7 shows a predistortion algorithm of a signal transmitted by the first unit according to a second embodiment of the invention.

DESCRIPTION OF THE INVENTION

[0011] Figure 1 shows two units, U1 and U2, of a telecommunications system exchanging data through two channels, H1 and H2, respectively. For example, the channels are two physical channels or two radio channels that are distinct from each other, that is defined by different characteristics. As an example, the invention is described in the present application within the framework of a radiocommunications system, although the said invention may be adapted to all types of applications in which two units send data through two different channels respectively. The two radio channels, H1 and H2, are defined by two frequencies that are different from each other.

[0012] With respect to Figure 2, the first unit U1 comprises, in the transmission chain, a modulator 10, a predistorter 11, an up-converter 12, a transmitting amplifier 13 and a duplexer 14. In the reception chain, the said first unit U1 comprises the duplexer 14, a reception amplifier 20, a down-converter 21, a demodulator 22 and a generator of predistortion coefficients 23.

[0013] Figure 4 shows a predistorter 11 of length N in accordance with an embodiment of the present invention. This predistorter has the structure of a conventional lineal equalizer, but all other types of predistorter may be used in accordance with the invention. This predistorter 11 carries out a function of "equalization" of the transmitted signal instead of carrying out an equalization of the signal received in the remote unit. The predistorter 11 comprises taps separated by time intervals, T/2. The signal $x(n)$ produced by the modulator 10 is applied to an input of the predistorter which delays the same through delay lines, T/2, connected in cascade. The output of the first delay line is $x(n-1)$, the output of the second delay line is $x(n-2)$, the output of the third delay line is $x(n-3)$, and so on repeatedly in such a way that the output of the last delay line is $x(n-N+1)$. There is a total of (N-1) delay lines and N predistortion coefficients. The N predistortion coefficients $c_0, c_1, c_2, \dots, c_{N-1}$ are complex numbers. Multipliers multiply the signal $x(n)$ and the different delayed signals $x(n-1), x(n-2), x(n-3), \dots, x(n-N+1)$ by the predistortion coefficients $c_0, c_1, c_2, \dots, c_{N-1}$, respectively. An adder adds the resulting multiplied signals to produce the predistorted signal which is applied to an input of the up-converter 12. In practice, the predistortion coefficients $c_0, c_1, c_2, \dots, c_{N-1}$ change periodically in accordance with the algorithms of Figures 6 and 7 in order to adapt to the characteristics of the channel H1.

[0014] The generator of predistortion coefficients 23 shown in figure 5 comprises a translator 230, a frequency response generator 231, an inverse translator 232, a division circuit of predistortion coefficients 234 and, alternatively, a coefficient selector 233. The output of the division circuit of predistortion coefficients 234 supplies the predistorter 11 with the appropriate predistortion coefficients $c_0, c_1, c_2, \dots, c_{N-1}$. The generator of

predistortion coefficients 23 may use a processor, a microprocessor or another signal treatment device.

[0015] A training signal, or sequence, received from the output of the demodulator 22 and sampled by a sampler (not shown) is applied to an input of the generator of predistortion coefficients 23. The sampled signal is received by the translator 230 which converts the temporal signal received into an equivalent signal in the frequency domain. For example, a DFT circuit (Discrete Fourier Transform) is used to provide the function of the translator 230. The resulting signal is applied to an input of the frequency response generator, or channel inverter, 231 which applies the next existing lineal relations in order to generate inverse coefficients: a received training signal, divided by a training signal as it really is, represents the channel transfer function, and the inverse of this transfer function corresponds to the division of the training signal as it really is and the received training signal.

[0016] In the present description, the terms "inverse coefficients" refer to coefficients which, in the temporal domain, when convoluted with the direct coefficients (that is, the coefficients that define the channel) give the Dirac impulse as a result, and, in the frequency domain, when multiplied with the direct coefficients, give the Fourier Transform of this impulse as a result. In other words, these inverse coefficients are the coefficients of the inverse channel.

[0017] The training signal received by the unit U1 is normally different from the training signal initially transmitted, and corresponds to a distorted version of the said transmitted signal. The training signal as it is really is stored locally in the frequency response generator 231.

[0018] The inverse translator 232 generates a temporal pulse response sequence $\{y(0), y(1), y(2), \dots, y(M-1)\}$ of the inverse channel. This temporal pulse response represents the inverse temporal characteristics of a channel, and defines a characteristic of finite pulse response of the said channel. In theory, if the finite pulse response sequence and the channel are in cascade, the overall pulse frequency response is equivalent to a bandpass filter with a pass band equal to the pass band of the channel.

[0019] The pulse response sequence $\{y(0), y(1), y(2), \dots, y(M-1)\}$ defines predistortion coefficients used by the division module of predistortion coefficients 234 to supply the coefficients of the predistorter of figure 4, as described below with respect to Figures 6 and 7. In this case, M is equal to N.

[0020] Alternatively, the predistorter 11 may use less coefficients (N coefficients) than the number M of coefficients $\{y(0), y(1), y(2), \dots, y(M-1)\}$ produced by the inverse translator 232. Accordingly, the selector 233 selects, for example, the coefficients associated with maximum energy of the received signal. These selected coefficients are applied to the division module of predistortion coefficients 234 to define the predistortion coeffi-

cients of the signal to be transmitted by the unit U1. An embodiment of a generator of predistortion coefficients 23 is described in US-A-5636244, starting from column 5, line 14, which, being referred to in the present application, is included in it.

[0021] With respect to Figure 6, an algorithm is now described for carrying out a method in accordance with a first alternative solution of the present invention. This algorithm is implemented partly in the coefficient division circuit 234 of the first unit U1 and partly in the second unit, or remote unit, U2.

[0022] Figure 6 shows a first embodiment of the present invention. In accordance with STEP 50, the first unit U1 sends a first predetermined sequence, or training sequence, $s(t)$ to the second unit U2 which sends it back in the form received to the first unit U1, in the form of a return sequence. With respect to figure 3, the second unit U2 comprises, in the transmission chain, a modulator 30, an up-converter 31, a transmitting amplifier 32 and a duplexer 50. In the reception chain, this second unit U2 comprises the duplexer 50, a reception amplifier 40, a down-converter 41, and a demodulator 42. Also, the second unit U2 comprises a timing circuit 34 whose input is connected to an output of the demodulator 42 as well as a switch 35 controlled by the timing circuit 34. The switch 35 is connected between an output of the down-converter 41 and an input of the up-converter 32. So that the second unit U2 may send back the sequence $s(t)$ received in the form received to the first unit U1, the timing circuit 34 detects the temporal location of the said sequence $s(t)$ and controls the switch 35 in such a way that, during the time the said sequence appears, the output of the down-converter 41 connects to an input of the up-converter 32. At the same time, the timing circuit 34 blocks the output of the modulator 30. This has the result that the sequence $s(t)$ sent by the first unit U1 is sent back in the form received by the second unit U2 to the first unit U1.

[0023] The first unit U1 receives the return signal $H1^*H2^*s(t)$, $H1$ being the transfer function of the associated channel H1, and $H2$ being the transfer function of the associated channel H2. In order to simplify the reference signs in the present specification, the terms $H1$ and $H2$ are used to designate channels, transfer functions or coefficients associated with these functions. In accordance with STEP 51, the first unit U1 computes representative coefficients of the first and second channels jointly on the basis of the return signal received.

[0024] In accordance with STEP 53, the first unit U1 also computes representative coefficients of the second channel H2 on the basis of a second predetermined sequence, or training sequence, $s'(t)$ (which may be the same as the sequence $s(t)$) sent (STEP 52) from the second unit U2 to the first unit U1. In accordance with STEP 54, according to these representative coefficients of the second channel H2 and of the representative coefficients of the first and second channels H1 and H2, the first unit U1 predistorts the signal sent to the second

unit U_2 with inverse coefficients H_1^{-1} of the first channel.

[0025] In accordance with a first sub-alternative of this first embodiment, the representative coefficients of the second channel are inverse coefficients $H2^{-1}$ of the second channel and the representative coefficients of the first and second channels jointly are direct coefficients $H1 \cdot H2$ of the first and second channels jointly. In this case, predistortion of the signal $S(t)$ sent by U1 is carried out with the inverse coefficients $H1^{-1}$ of the convolution of the inverse coefficients of the second channel $H2^{-1}$ and the direct coefficients of the first and second channels $H1 \cdot H2$. The generator of predistortion coefficients 23 produces inverse coefficients according to the signal received and a predetermined (training) sequence stored locally. In order to produce direct coefficients from the inverse coefficients received from the generator 23, the coefficient division circuit 234 includes a coefficient inversion circuit of a type known in the state of the art.

[0026] In accordance with a second sub-alternative of the first embodiment illustrated in detail in Figure 6, the representative coefficients of the second channel H2 are direct coefficients of the second channel and the representative coefficients of the first and second channels jointly are inverse coefficients $(H1^*H2)^{-1}$ of the first and second channels jointly. In this case, predistortion is carried out with the coefficients $H1^{-1}$ of the convolution of the direct coefficients of the second channel and the inverse coefficients of the first and second channels jointly. As indicated above, the generator of predistor-
tion coefficients 23 produces inverse coefficients according to the signal received and a predetermined sequence stored locally. In order to produce direct coefficients from the inverse coefficients received from the generator 23, the coefficient division circuit 234 includes a coefficient inversion circuit of a type known in the state of the art.

[0027] For this first embodiment, in the first unit U1, the circuitry 23 computes the representative coefficients of the second channel H2 or $H2^{-1}$ on the basis of the predetermined sequence $s'(t)$ sent from the remote unit U2. The transmission chain of the first unit U1 sends the first predetermined sequence $s(t)$ to the remote unit U2 which sends it back in the form received to the unit U1, as described with respect to figure 3. The circuitry 23 adjusts the representative coefficients $H1^*H2$ or $(H1^*H2)^{-1}$ of the first and second channels jointly on the basis of the return sequence received. The predistorter 11 predistorts the sent signal $S(t)$ with the inverse coefficients $H1^{-1}$ of the first channel which are computed according to the representative coefficients of the second channel and the representative coefficients of the first and second channels jointly.

[0028] In the first sub-alternative, the circuitry computes the inverse coefficients H_2^{-1} of the second channel and computes the direct coefficients $H_1 \cdot H_2$ of the first and second channels jointly. In this case, the pre-

distorter 11 predistorts the signal to be sent by means of the inverse coefficients $H1^{-1}$ of the convolution of the inverse coefficients of the second channel $H2^{-1}$ and of the direct coefficients of the first and second channels $H1 \cdot H2$. These inverse coefficients $H1^{-1}$ of the convolution of the inverse coefficients of the second channel $H2^{-1}$ and of the direct coefficients of the first and second channels $H1 \cdot H2$ are computed in the coefficient division circuit 234.

10 [0029] In the second sub-alternative, the circuitry 23 computes the direct coefficients $H2$ of the second channel and computes the inverse coefficients $(H1^*H2)^{-1}$ of the first and second channels jointly. In this case, the predistorter 11 predistorts the signal to be sent by
15 means of the coefficients $H1^{-1}$ of the convolution of the

[0030] Figure 7 shows a second embodiment of the present invention.

[0031] In accordance with this second embodiment, the first unit U1 computes representative coefficients H2⁻¹ of the second channel (STEP 61) on the basis of a second sequence, typically a training sequence, s'(t) sent from the second unit U2 (STEP 60).

25 [0032] Also, the first unit U1 sends to the second unit U2 (STEP 62) a first sequence $H2^{-1}s(t)$ which results from the convolution of a predetermined sequence $s(t)$ and of the said representative coefficients $H2^{-1}$ of the

second channel, the said first sequence being sent back as received by the second unit U2 to the first unit U1 in the form of a return sequence. The first unit U1 receives the signal $H2^{-1}s(t)*H1*H2 = H1*s(t)$. The first unit U1 computes representative coefficients of the first channel $H1^{-1}$ on the basis of the return sequence, and predistorts the signal sent to the second unit U2 according to these representative coefficients of the first channel. The representative coefficients of the second channel

The representative coefficients of the second channel are inverse coefficients $H2^{-1}$ of the second channel and the representative coefficients $H1^{-1}$ of the first channel are inverse coefficients of the convolution of the said return sequence $H2^{-1} * H2^* H1^{-1} s(t)$ received by the first unit $U1$ and of an inverse predetermined sequence $s^{-1}(t)$.

[0033] In accordance with this second embodiment, in the unit U1, the circuitry 23 computes the inverse coefficients H_2^{-1} of the second channel on the basis of the second sequence $s'(t)$ sent from the second unit U2 to the first unit U1. The transmission chain of this first unit

50 U1 sends to the second unit U2 a first sequence $H2^{-1}s(t)$ which results from the convolution of a predetermined sequence $s(t)$ and of the inverse coefficients $H2^{-1}$ of the second channel, the said convoluted first sequence being sent back as received by the second unit U2 to the first unit U1 in the form of a return sequence $H2^{-1}H1^{-1}H2^{-1}s(t)$. The circuitry 23 computes the inverse coefficients of the first channel $H1^{-1}$ on the basis of this return sequence.

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[0034] In practice, the predistortion coefficients are

periodically brought up to date by means of the algorithms of figures 6 and 7 which are repeated over and over again.

[0035] Moreover, although the terms "predetermined sequence" or "training sequence" are used in the present specification, these terms cover any signal that allows the inverse coefficients $H1^{-1}$ of the first channel to be determined, using, for example, the process of blind equalization to determine the characteristic coefficients of the channels $H1$, $H1^*H2$, etc... 5

Claims

1. Method for predistorting a signal sent from a first unit (U1) to a second unit (U2), the said first unit (U1) sending data to the said second unit (U2) through a first channel (H1) and the said second unit (U2) sending data to the said first unit through a second channel (H2), the said method comprising the following step: 15
 - in the first unit, the computing of representative coefficients of the second channel (H2) on the basis of a second predetermined sequence ($s'(t)$) sent from the second unit (U2) to the first unit (U1); characterized in that it comprises the following steps: 20
 - the sending by the first unit (U1) of a first predetermined sequence ($s(t)$) to the second unit (U2) which sends it back in the form received to the first unit (U1), in the form of a return sequence; 25
 - in the said first unit (U1), the computing of representative coefficients of the first and second channels jointly ($H1^*H2$; $(H1^*H2)^{-1}$), on the basis of the return sequence received ($H1^*H2^*s(t)$); and 30
 - predistortion of the signal sent by the first unit by means of inverse coefficients ($H1^{-1}$) of the first channel which are computed according to the said representative coefficients of the second channel ($H2$, $H2^{-1}$) and the said representative coefficients of the first and second channels jointly ($H1^*H2$, $(H1^*H2)^{-1}$). 35
2. Method according to claim 1, characterized in that the said representative coefficients of the second channel are inverse coefficients ($H2^{-1}$) of the second channel, and the said representative coefficients of the first and second channels jointly are direct coefficients ($H1^*H2$) of the first and second channels jointly, the said predistortion being carried out with the inverse coefficients ($H1^{-1}$) of the convolution of both the inverse coefficients of the second channel ($H2^{-1}$) and the direct coefficients of the first and second channels jointly ($H1^*H2$). 40
3. Method according to claim 1, characterized in that 45

the said representative coefficients of the second channel (H2) are direct coefficients (H2) of the second channel and the said representative coefficients of the first and second channels jointly are inverse coefficients ($H1^*H2^{-1}$) of the first and second channels jointly, the said predistortion being carried out by means of the coefficients ($H1^{-1}$) of the convolution of the direct coefficients of the second channel and of the inverse coefficients of the first and second channels. 50

4. Method for predistorting a signal sent from a first unit (U1) to a second unit (U2), the said first unit (U1) sending data to the said second unit (U2) through a first channel (H1) and the said second unit (U2) sending data to the said first unit (U1) through a second channel (H2), the said method comprising the following step:
 - in the first unit, the computing of representative coefficients of the second channel ($H2$, $H2^{-1}$) on the basis of a second predetermined sequence ($s'(t)$) sent from the second unit (U2) to the first unit (U1); characterized in that it comprises the following steps:
 - the sending by the first unit (U1) to the second unit (U2) of a first sequence ($H2^{-1}s(t)$) which results from the convolution of a predetermined sequence ($s(t)$) and of the said representative coefficients ($H2^{-1}$) of the second channel, the said first sequence being sent back as received by the second unit to the first unit (U1), in the form of a return sequence ($H1^*H2^*H2^{-1}s(t)$);
 - in the said first unit (U1), the computing of inverse coefficients of the first channel ($H1^{-1}$) on the basis of the said return sequence; and
 - predistortion of the signal sent by the first unit (U1) on the basis of the said inverse coefficients of the first channel. 55
5. Method according to claim 4, characterized in that the said representative coefficients of the second channel are inverse coefficients ($H2^{-1}$) of the second channel (U2), and the said representative coefficients ($H1^{-1}$) of the first channel (U1) are inverse coefficients of the convolution of the said return sequence ($H1^*H2^*H2^{-1}s(t)$) received by the first unit (U1) and of an inverse predetermined sequence ($s^{-1}(t)$). 60
6. Unit (U1) of a telecommunications system that predistsorts a signal to be sent to a remote unit (U2), comprising:
 - means (11) to predistort the said signal ($S(t)$) to be sent;
characterized in that it comprises:
 - means (53) to compute representative coeffi- 65

cients of the second channel (H_2) on the basis of a second predetermined sequence ($s(t)$) sent (52) from the remote unit (U2) to the unit (U1);

- means (50) to send a first predetermined sequence ($s(t)$) to the remote unit (U2), which sends it back in the form received to the said unit (U1), in the form of a return sequence;
- means (51) to compute representative coefficients of the first and second channels jointly (H_1, H_2), on the basis of the return sequence received ($s(t)$); and in that
- the said predistorting means (11) predistort the signal sent ($S(t)$) by means of the inverse coefficients of the first channel which are computed according to the said representative coefficients of the second channel (H_2, H_2^{-1}) and the said representative coefficients of the first and second channels jointly ($\{H_1 \cdot H_2\}^{-1}, \{H_1 \cdot H_2\}$).

7. Remote unit for sending back to the unit (U1) according to claim 6, in the form as received, the said first predetermined sequence ($s(t)$) in the form of a return sequence, characterized in that it comprises:

- means to connect, in a way that is synchronized with the arrival of the predetermined sequence ($s(t)$), an output of a down-converter with an input of an up-converter.

8. Unit according to claim 6, characterized in that the means (53) for computing representative coefficients of the second channel (H_2) compute inverse coefficients (H_2^{-1}) of the second channel, the means (51) for computing representative coefficients of the first and second channels jointly (H_1, H_2) compute direct coefficients ($H_1 \cdot H_2$) of the first and second channels jointly, and the predistorting means (11) predistort the signal to be sent by means of the inverse coefficients (H_1^{-1}) of the convolution of inverse coefficients of the second channel (H_2^{-1}) and direct coefficients of the first and second channels jointly ($H_1 \cdot H_2$).

9. Unit according to claim 6, characterized in that the means (53) for computing representative coefficients of the second channel (H_2) compute direct coefficients (H_2) of the second channel, the means (51) for computing representative coefficients of the first and second channels jointly (H_1, H_2) compute inverse coefficients ($(H_1 \cdot H_2)^{-1}$) of the first and second channels jointly, and the predistorting means (11) predistort the signal to be sent by means of the coefficients (H_1^{-1}) of the convolution of the direct coefficients of the second channel and of the inverse coefficients of the first and second channels jointly.

10. Unit (U1) of a telecommunications system that predistorts a signal to be sent to a second unit (U2), comprising:

- means (11) for predistorting the said signal ($S(t)$) to be sent; characterized in that it comprises
- means (61) to compute inverse coefficients (H_2^{-1}) of the second channel (U2) on the basis of a second sequence ($s'(t)$) sent from the second unit (U2) to the first unit (U1);
- means (62) to send to the second unit (U2) a first sequence ($H_2^{-1} \cdot s(t)$) which results from the convolution of a predetermined sequence ($s(t)$) and of the said inverse coefficients (H_2^{-1}) of the second channel (H_2), the said first sequence being sent back as received by the second unit (U2) to the first unit (U1), in the form of a return sequence ($H_2^{-1} \cdot H_1 \cdot H_2 \cdot s(t)$);
- means to compute inverse coefficients of the first channel (H_1^{-1}) on the basis of the said return sequence; and
- the said predistorting means predistort the signal to be sent (U1) according to the said inverse coefficients of the first channel.

11. Remote unit for sending back to the unit (U1) according to claim 10, in the form as received, the said first sequence ($H_2^{-1} \cdot s(t)$) in the form of a return sequence, characterized in that it comprises:

- means to connect, in a way that is synchronized with the arrival of the first sequence ($s(t)$), an output of a down-converter with an input of an up-converter.

FIG 1

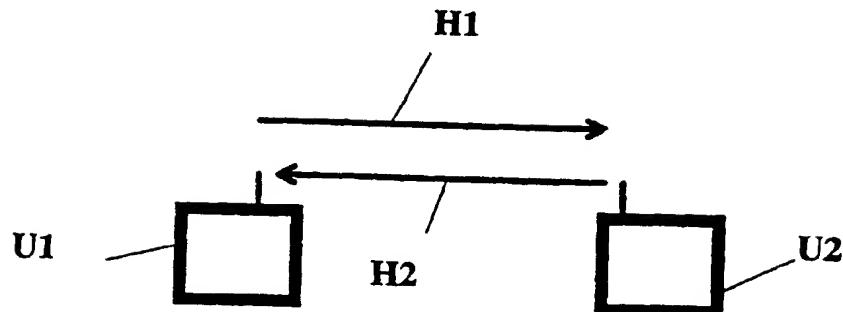


FIG 2

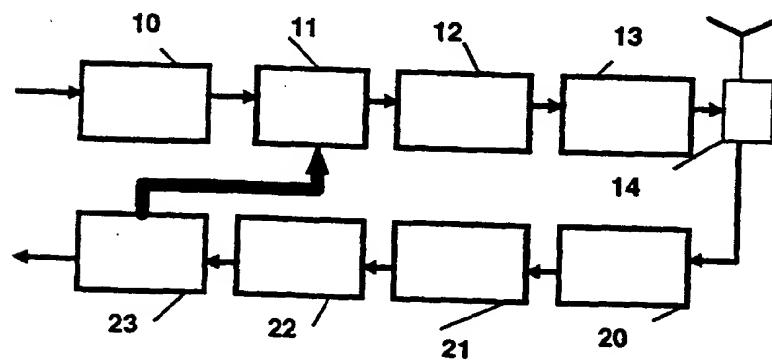


FIG 3

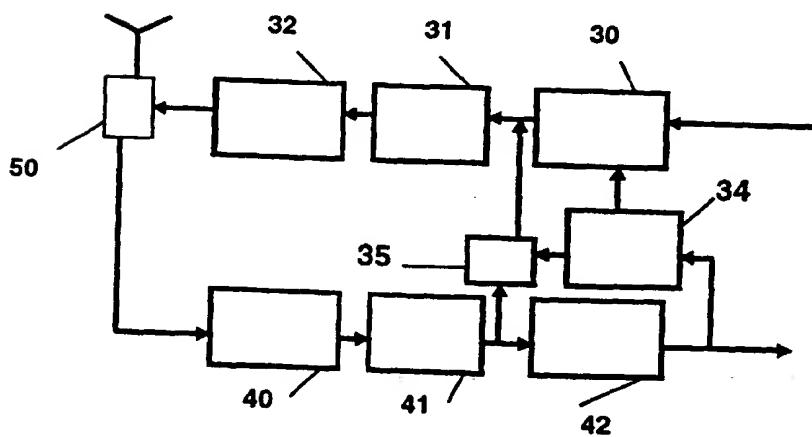


FIG 4

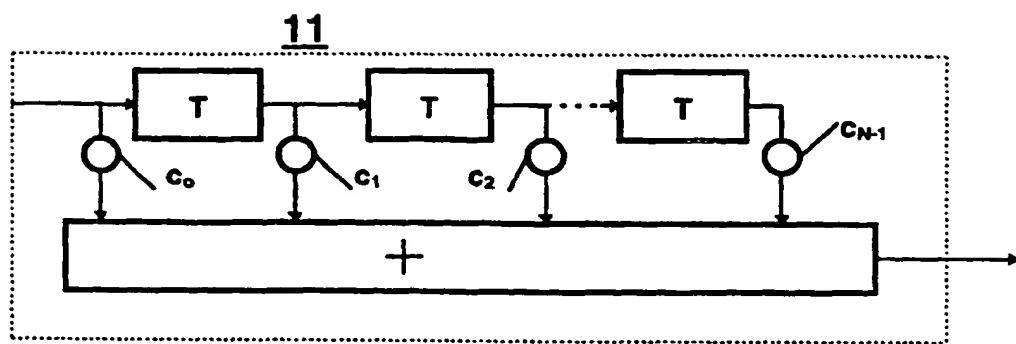


FIG 5

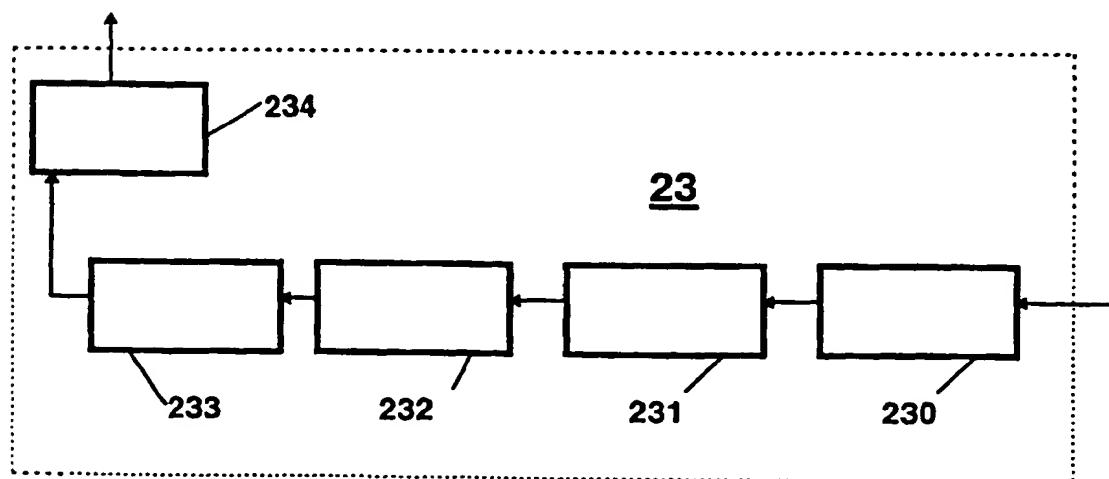


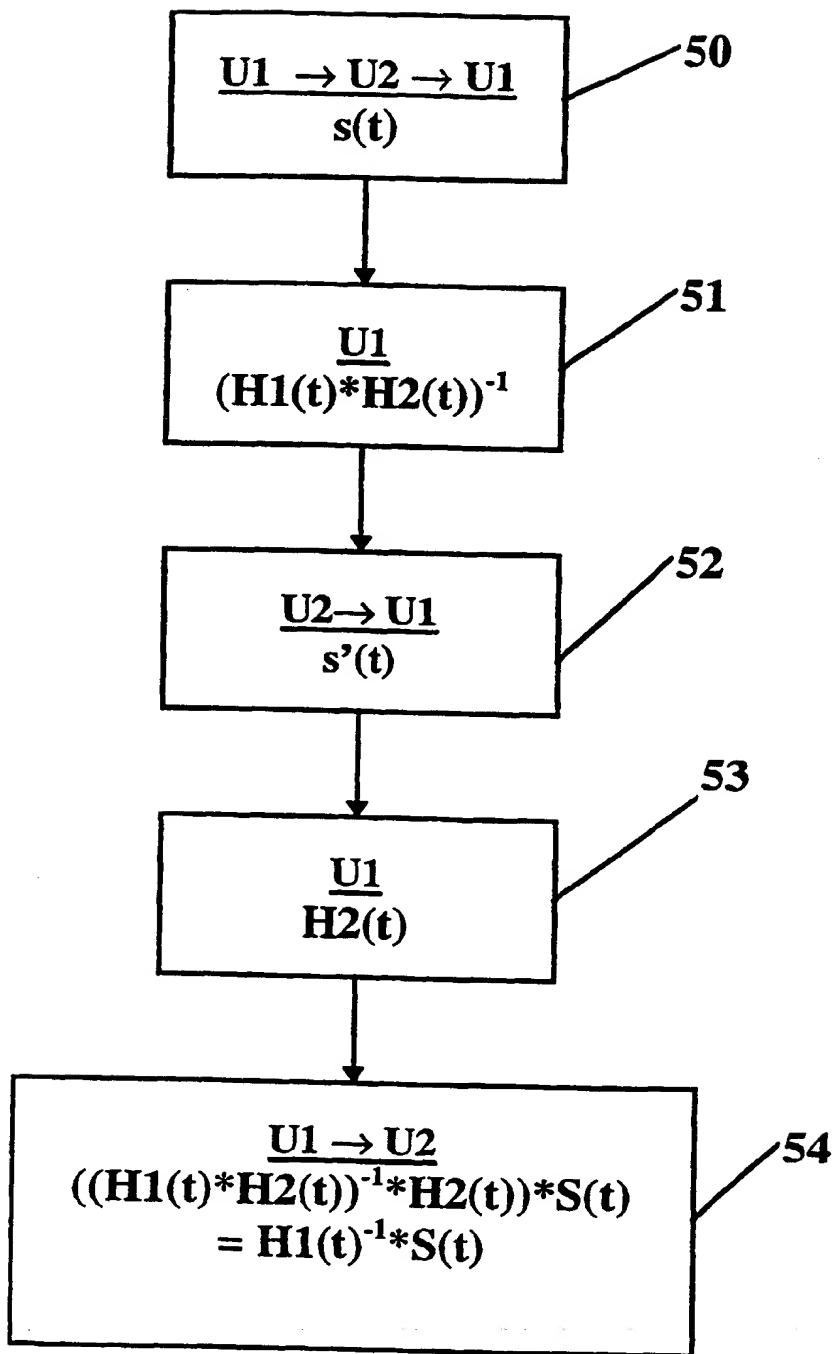
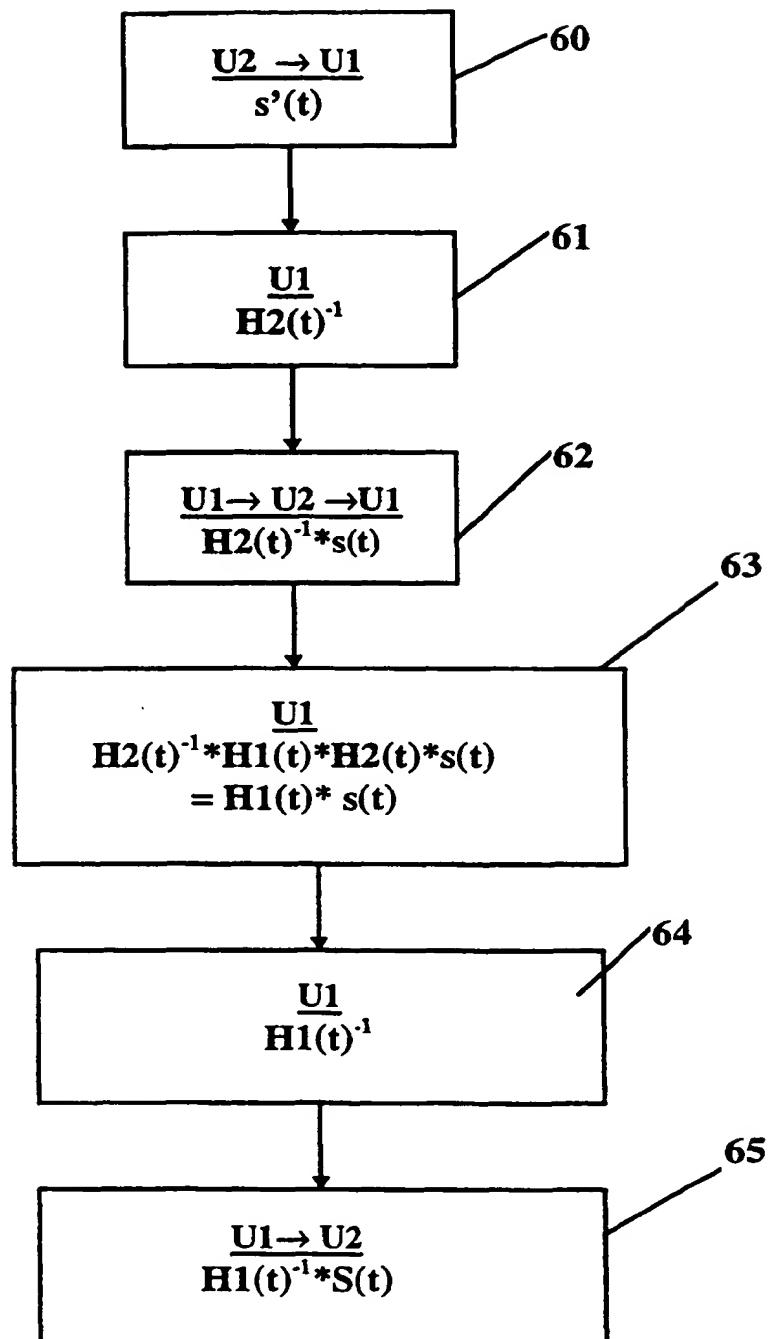
FIG 6

FIG 7



European Patent
Office

EUROPEAN SEARCH REPORT

Application Number

EP 98 50 0045

DOCUMENTS CONSIDERED TO BE RELEVANT									
Category	Citation of document with indication, where appropriate, of relevant passages	Relevant to claim	CLASSIFICATION OF THE APPLICATION (Int.Cl.6)						
A	<p>AL-MASHOUQ ET AL.: "Adaptive pre-equalization using neural-like algorithm" 1996 IEEE DIGITAL SIGNAL PROCESSING WORKSHOP, 1 - 4 September 1996, pages 439-441, XP002073086 New York, US * page 439, right-hand column, paragraph 3 *</p> <p>—</p> <p>SESAV, GIBBARD: "Asymmetric signal processing for indoor wireless LANs" PROCEEDINGS OF THE VEHICULAR TECHNOLOGY CONFERENCE, 25 - 28 July 1995, pages 714-718, XP000551627 New York, US * page 714, right-hand column, paragraph 3 * * page 714, right-hand column, paragraph 4 *</p> <p>—</p> <p>EP 0 813 300 A (LORAL AEROSPACE) 17 December 1997 * column 2, line 24 - line 39 *</p> <p>—</p>	1,4,6,10	H04L25/02 H04L5/14						
			TECHNICAL FIELDS SEARCHED (Int.Cl.6)						
			H04L						
<p>The present search report has been drawn up for all claims</p> <table border="1" style="width: 100%; border-collapse: collapse;"> <tr> <td style="width: 33%;">Place of search</td> <td style="width: 33%;">Date of completion of the search</td> <td style="width: 34%;">Examiner</td> </tr> <tr> <td>THE HAGUE</td> <td>29 July 1998</td> <td>Scriven, P</td> </tr> </table>				Place of search	Date of completion of the search	Examiner	THE HAGUE	29 July 1998	Scriven, P
Place of search	Date of completion of the search	Examiner							
THE HAGUE	29 July 1998	Scriven, P							
EPO FORM 1603 OJ/A2 (P04C01) CATEGORY OF CITED DOCUMENTS X : particularly relevant if taken alone Y : particularly relevant if combined with another document of the same category A : technological background O : non-written disclosure P : intermediate document		T : theory or principle underlying the invention E : earlier patent document, but published on, or after the filing date D : document cited in the application L : document cited for other reasons & : member of the same patent family, corresponding document							

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ON EUROPEAN PATENT APPLICATION NO.

EP 98 50 0045

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29-07-1998

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